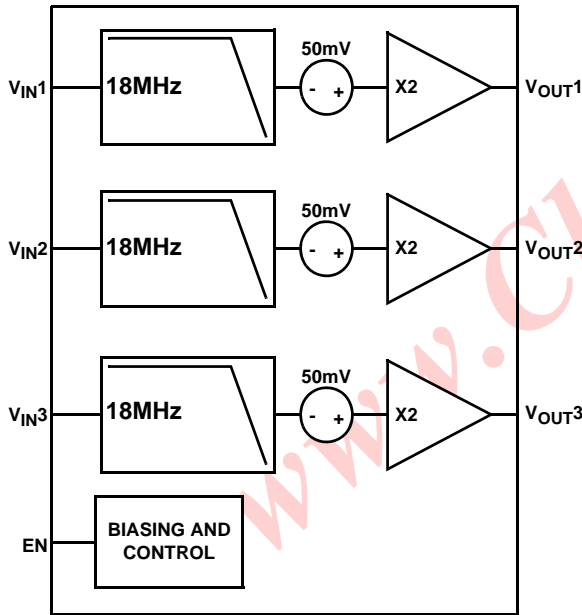


Triple Channel Video Driver with LPF

The ISL59121 is a triple channel reconstruction filter with a -3dB roll-off frequency of 18MHz. Operating from single supplies ranging from +2.5V to +3.6V and sinking a low 4mA quiescent current, the ISL59121 is ideally suited for low power, battery-operated applications. An enable pin allows the part to be placed in a 500nA shutdown mode in less than 30ns.

The ISL59121 is designed to meet the needs for extremely low power and bandwidth requirements in battery-operated communication, instrumentation, and modern industrial applications such as video on demand, cable set-top boxes, MP3 players, and HDTV. The ISL59121 is offered in a space-saving chipscale package guaranteed to a 0.57mm maximum height constraint and specified for operation from -40°C to +85°C temperature range.

Block Diagram



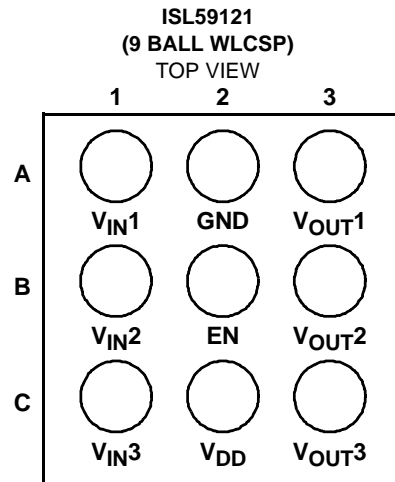
Features

- 3rd order 18MHz reconstruction filter
- 4mA typical supply current
- Less than 500nA maximum power-down current
- 2.5V to 3.6V supply range
- CSP package
- Pb-free plus anneal (RoHS compliant)

Applications

- Video amplifiers
- Portable and handheld products
- Communications devices
- Video on demand
- Cable set-top boxes
- Satellite set-top boxes
- MP3 players
- HDTV
- Personal video recorder

Pinout



Ordering Information

| PART NUMBER (Note) | PART MARKING | TAPE AND REEL | TEMP. RANGE (°C) | PACKAGE (Pb-free) | PKG. DWG. # |
|--------------------|--------------|---------------|------------------|-------------------|-------------|
| ISL59121IIZ-T7 | 121Z | 7" | -40 to +85 | 9 Ball 3x3 WLCSP | W3x3.9B |

NOTE: Intersil Pb-free plus anneal products employ special Pb-free material sets; molding compounds/die attach materials and 100% matte tin plate termination finish, which are RoHS compliant and compatible with both SnPb and Pb-free soldering operations. Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.

ISL59121

Absolute Maximum Ratings (T_A = +25°C)

| | |
|--|------------------------------------|
| Supply Voltage from V _{DD} to GND | 4.2V |
| Input Voltage | V _{DD} +0.3V to GND -0.3V |
| Continuous Output Current | 40mA |
| Storage Temperature | -65°C to +125°C |
| Ambient Operating Temperature | -40°C to +85°C |
| Operating Junction Temperature | +125°C |
| ESD Rating | |
| Human Body Model (Per MIL-STD-883 Method 3015.7) | 2500V |
| Machine Model (Per EIAJ ED-4701 Method C-111) | 300V |

Thermal Information

| | |
|------------------------------|------------------------|
| Thermal Resistance (Typical) | θ _{JA} (°C/W) |
| 9 Ball WLCSP | +105 |

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typ values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: T_J = T_C = T_A

Electrical Specifications V_{DD} = 3.3V, T_A = +25°C, R_{SOURCE} = 200Ω, R_L = 150Ω to GND, unless otherwise specified.

| PARAMETER | DESCRIPTION | CONDITIONS | MIN | TYP | MAX | UNIT |
|------------------------------|--|---|------|--------|-------|------|
| INPUT CHARACTERISTICS | | | | | | |
| V _{DD} | Supply Voltage Range | | 2.5 | | 3.6 | V |
| I _{DD} | Quiescent Supply Current | V _{IN} = 500mV, EN = V _{DD} , no load | | 4.0 | 6.0 | mA |
| I _{DD_OFF} | Shutdown Supply Current | EN = 0V | | | 0.5 | μA |
| C _{IN} | V _{IN} Input Capacitance | | | | 10 | pF |
| I _{IN} | V _{IN} Input Bias Current | 0.0V < V _{IN} < 2.0V | | 0.01 | 1 | μA |
| V _{OLS} | Output Level Shift Voltage | V _{IN} = 0V, no load (minimum output voltage) | 45 | 100 | 150 | mV |
| A _V | Voltage Gain | R _L = 150Ω | 1.95 | 1.99 | 2.04 | V/V |
| ΔA _V | Channel-to Channel Gain Mismatch | | | ±0.5 | ±1.75 | % |
| PSRR | DC Power Supply Rejection | V _{DD} = 2.7V to 3.3V | | 63 | | dB |
| V _{OH} | Output Voltage High Swing | V _{IN} = 2V, R _L = 150Ω to GND | 2.85 | 3.1 | | V |
| I _{SC} | Output Short-Circuit Current | V _{IN} = 2V, V _{OUT} shorted to GND through 10Ω | 100 | 140 | | mA |
| | | V _{IN} = 0V, V _{OUT} shorted to V _{DD} through 10Ω | 140 | 200 | | mA |
| I _{ENABLE} | Enable Input Current | 0V < V _{EN} < 3.3V | | ±0.003 | ±1 | μA |
| V _{IL} | Disable Threshold | V _{DD} = 2.7V to 3.3V | | | 0.8 | V |
| V _{IH} | Enable Threshold | V _{DD} = 2.7V to 3.3V | 2.0 | | | V |
| R _{OUT} | Shutdown Output Impedance | EN = 0V DC | 5.0 | 6.4 | 8.0 | kΩ |
| | | EN = 0V, f = 4.5MHz | | 1.5 | | kΩ |
| AC PERFORMANCE | | | | | | |
| BW | -3dB Bandwidth | R _L = 150Ω, C _L = 5pF | | 18 | | MHz |
| Passband Gain | Normalized Gain at 11MHz | R _L = 150Ω, C _L = 5pF | -0.5 | | +0.5 | dB |
| Stopband Gain | Normalized Stopband Gain (minimum numbers from simulation, not able to production test). | f = 43MHz | -11 | -16 | | dB |
| | | f = 54MHz | -16 | -21 | | dB |
| dG | Differential Gain | NTSC and PAL | | 0.10 | | % |
| dP | Differential Phase | NTSC and PAL | | 0.5 | | ° |
| D/DT | Group Delay Variation | f = 100kHz, 5MHz | | 5.4 | | ns |
| SNR | Signal To Noise Ratio | 100% white signal | | 65 | | dB |
| +SR | Positive Slew Rate | 10% to 90%, V _{IN} = 1V step | 40 | 55 | | V/μs |

ISL59121

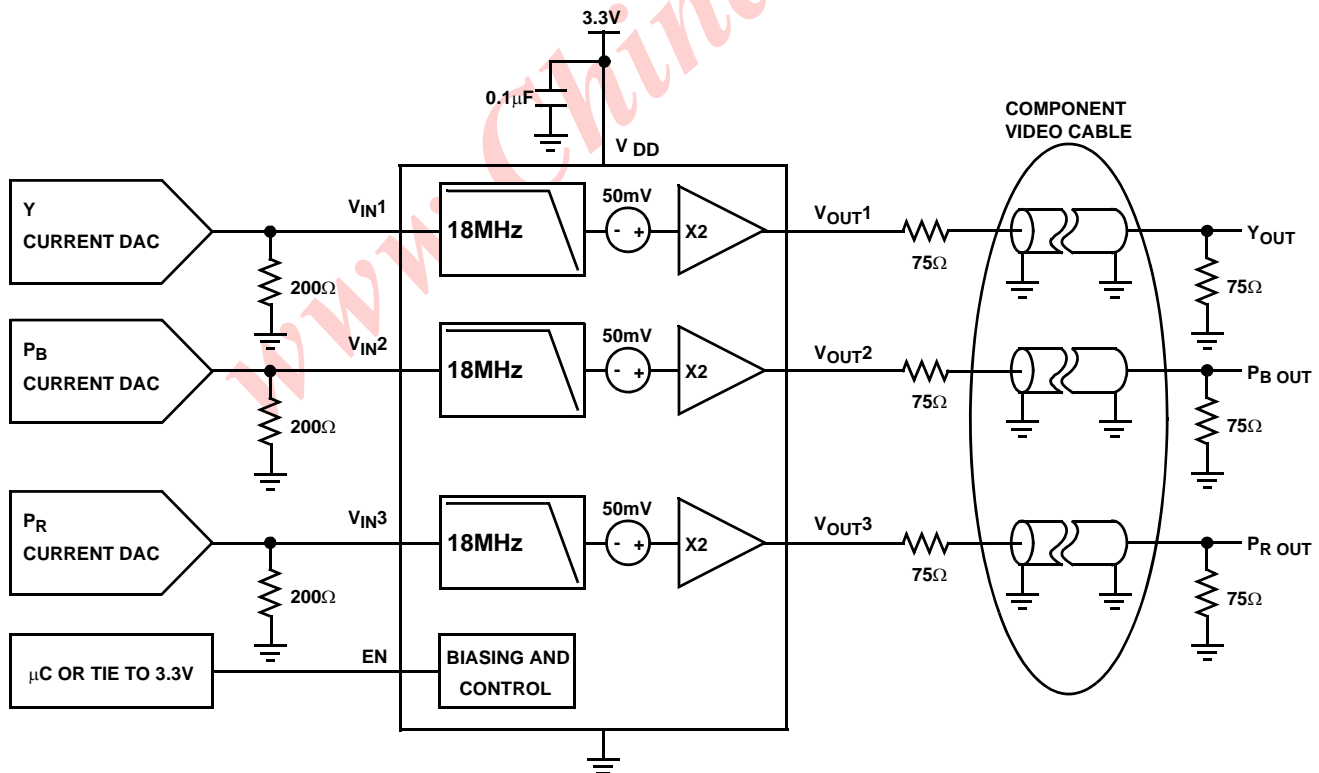
Electrical Specifications $V_{DD} = 3.3V$, $T_A = +25^\circ C$, $R_{SOURCE} = 200\Omega$, $R_L = 150\Omega$ to GND, unless otherwise specified. **(Continued)**

| PARAMETER | DESCRIPTION | CONDITIONS | MIN | TYP | MAX | UNIT |
|-----------|--------------------|------------------------------------|-----|-----|-----|------------|
| -SR | Negative Slew Rate | 90% to 10%, $V_{IN} = 1V$ step | 50 | 60 | | V/ μs |
| t_F | Fall Time | $2.5V_{STEP}$, 80% - 20% | | 25 | | ns |
| t_R | Rise Time | $2.5V_{STEP}$, 20% - 80% | | 22 | | ns |
| t_{ON} | Enable Time | $V_{IN} = 500mV$, V_{OUT} to 1% | | 250 | | ns |
| t_{OFF} | Disable Time | $V_{IN} = 500mV$, V_{OUT} to 1% | | 30 | | ns |

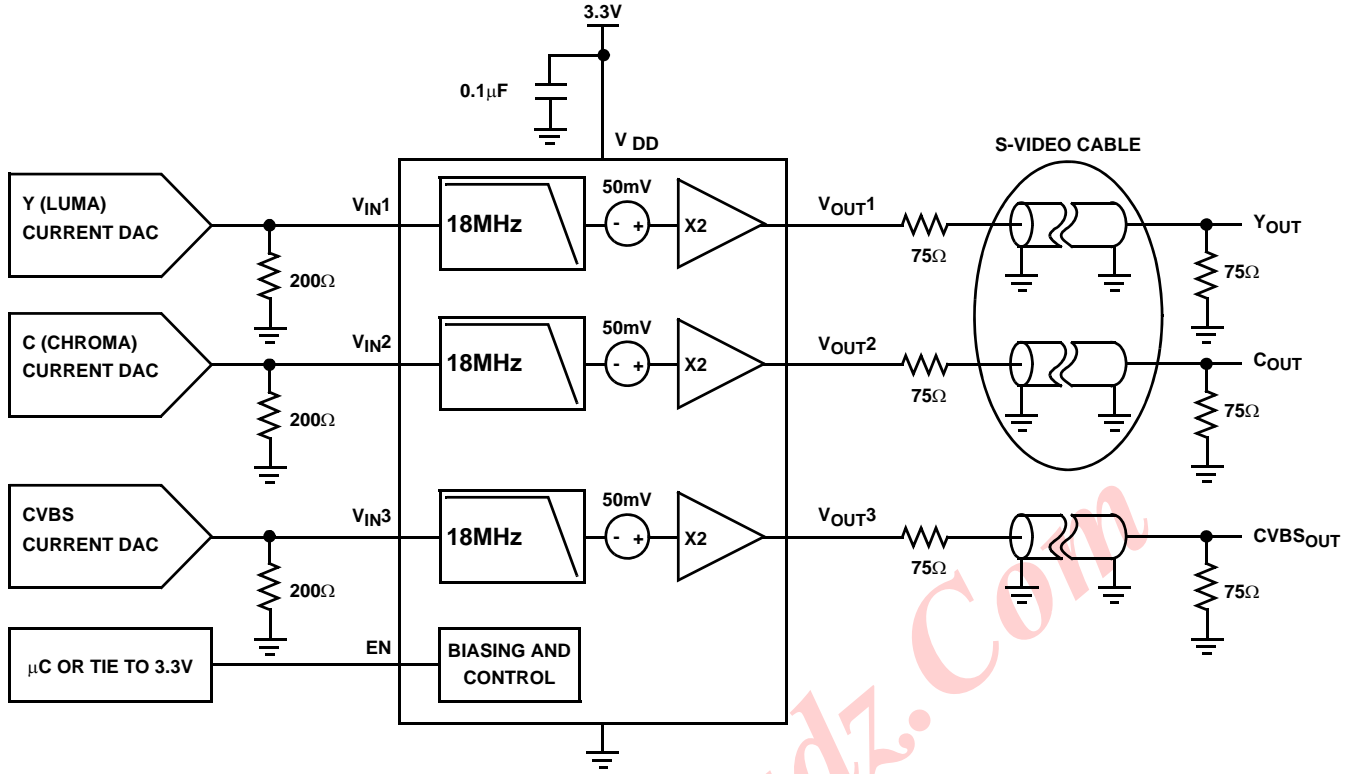
Pin Descriptions

| PIN NUMBER | PIN NAME | DESCRIPTION |
|------------|------------|----------------------------|
| A1 | V_{IN1} | Video input for Channel 1 |
| A2 | GND | Ground |
| A3 | V_{OUT1} | Video output for Channel 1 |
| B1 | V_{IN2} | Video input for Channel 2 |
| B2 | EN | Enable |
| B3 | V_{OUT2} | Video output for Channel 2 |
| C1 | V_{IN3} | Video input for Channel 3 |
| C2 | V_{DD} | Positive power supply |
| C3 | V_{OUT3} | Video output for Channel 3 |

Component Video Connection Diagram



Composite and S-Video Connection Diagram



Typical Performance Curves $V_{DD} = +3.3V$, $R_L = 150\Omega$, unless otherwise noted

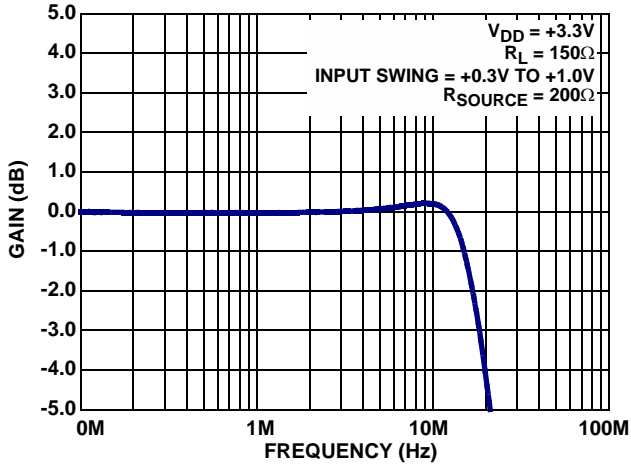


FIGURE 1. GAIN vs FREQUENCY -0.1dB

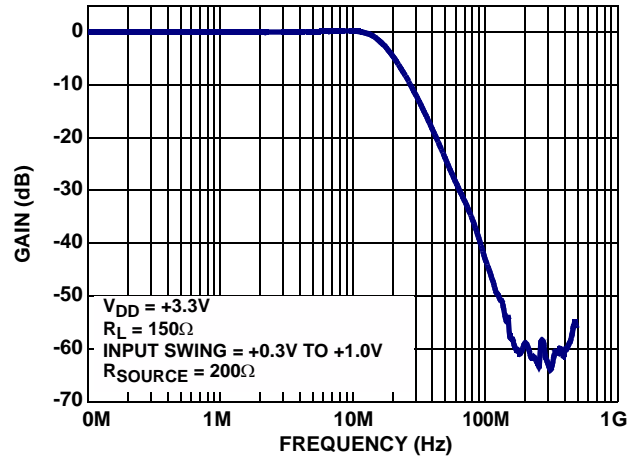


FIGURE 2. GAIN vs FREQUENCY -3dB POINT

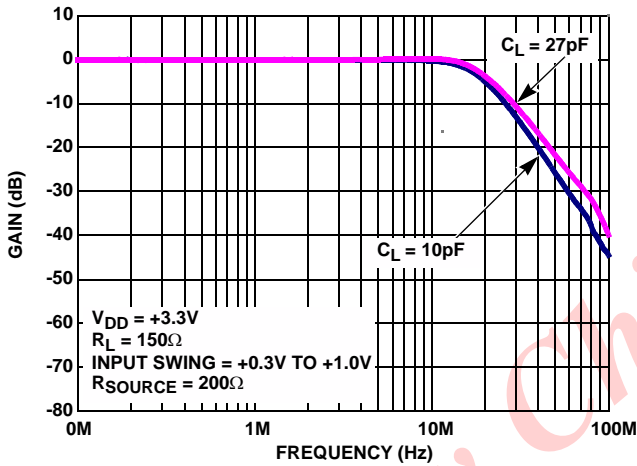


FIGURE 3. GAIN vs FREQUENCY FOR VARIOUS C_{LOAD}

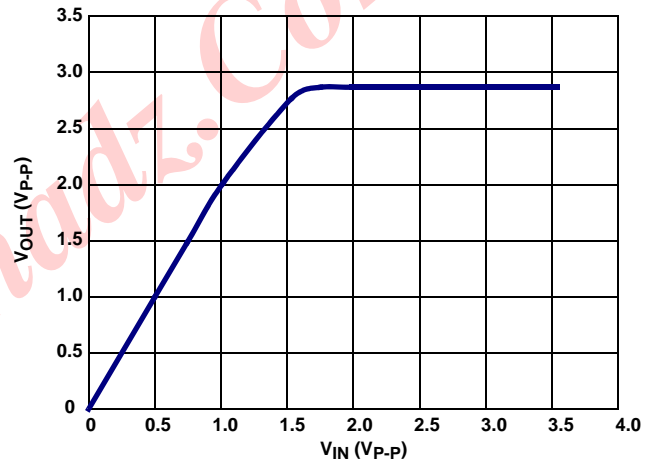


FIGURE 4. MAXIMUM OUTPUT MAGNITUDE vs INPUT MAGNITUDE

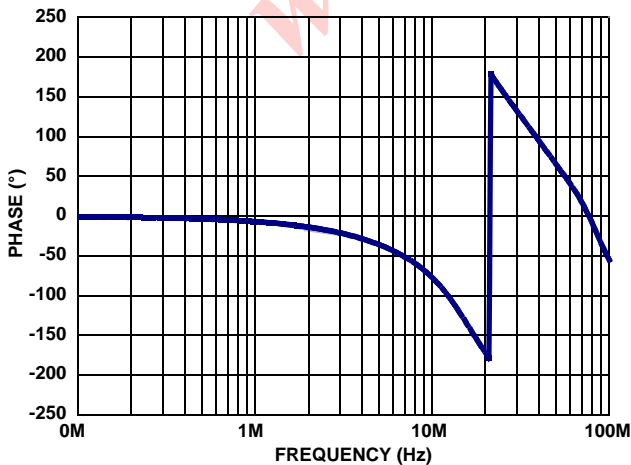


FIGURE 5. PHASE vs FREQUENCY

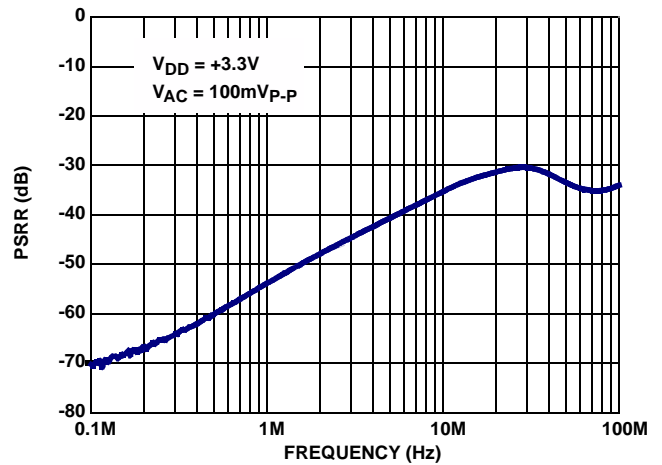


FIGURE 6. PSRR vs FREQUENCY

Typical Performance Curves $V_{DD} = +3.3V$, $R_L = 150\Omega$, unless otherwise noted

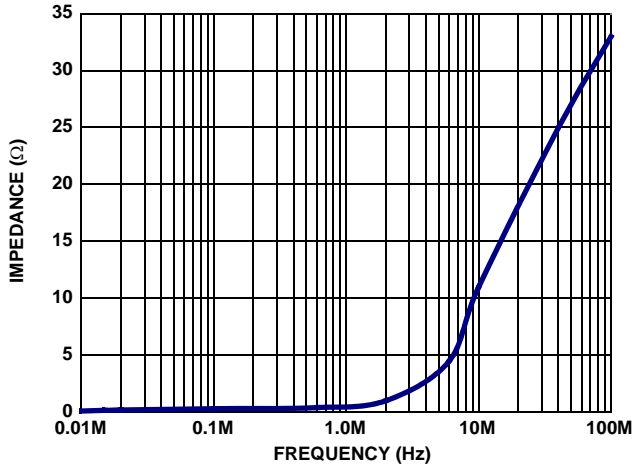


FIGURE 7. OUTPUT IMPEDANCE vs FREQUENCY

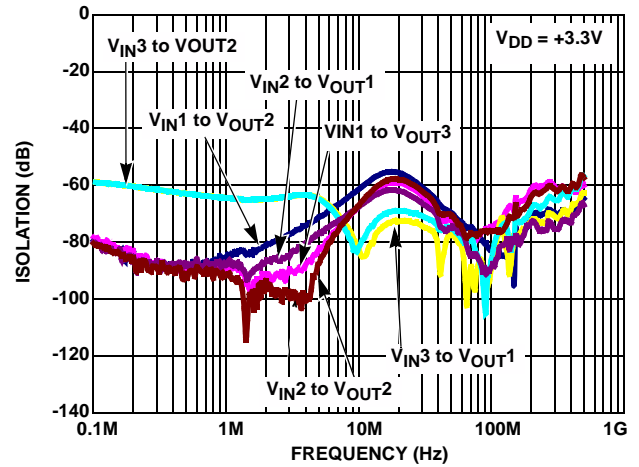


FIGURE 8. ISOLATION vs FREQUENCY

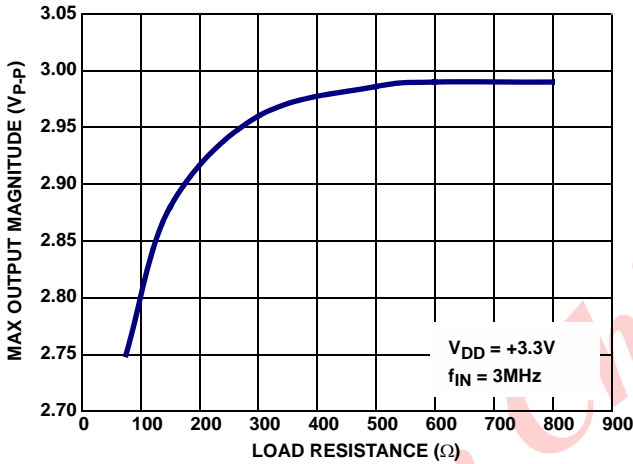


FIGURE 9. MAXIMUM OUTPUT vs LOAD RESISTANCE

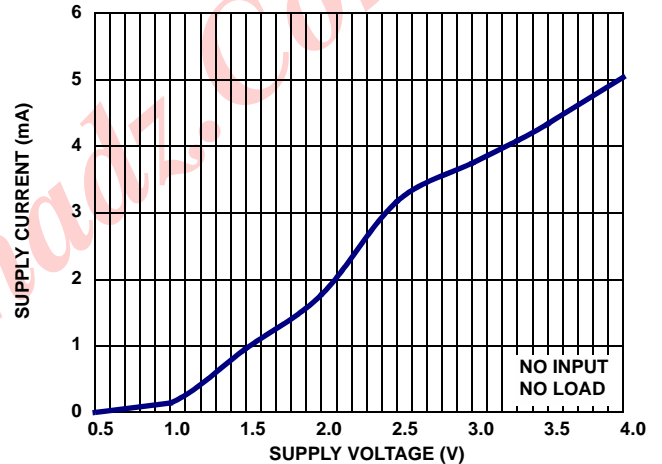


FIGURE 10. SUPPLY CURRENT vs SUPPLY VOLTAGE

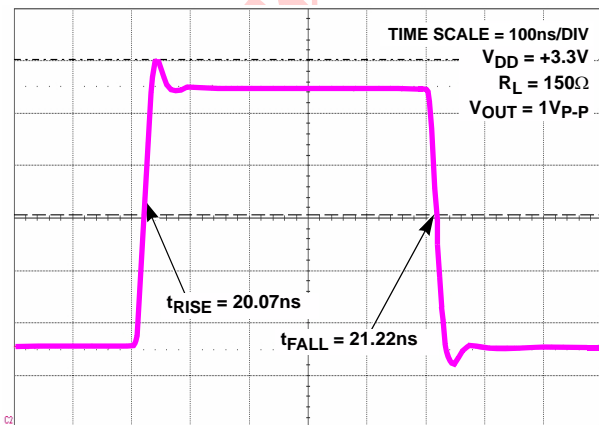


FIGURE 11. LARGE SIGNAL STEP RESPONSE

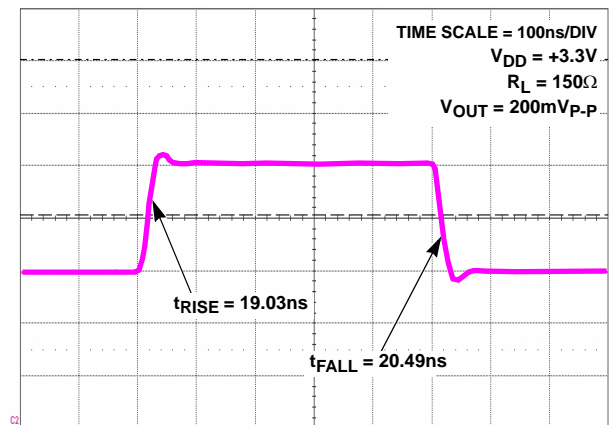


FIGURE 12. SMALL SIGNAL STEP RESPONSE

Typical Performance Curves $V_{DD} = +3.3V$, $R_L = 150\Omega$, unless otherwise noted

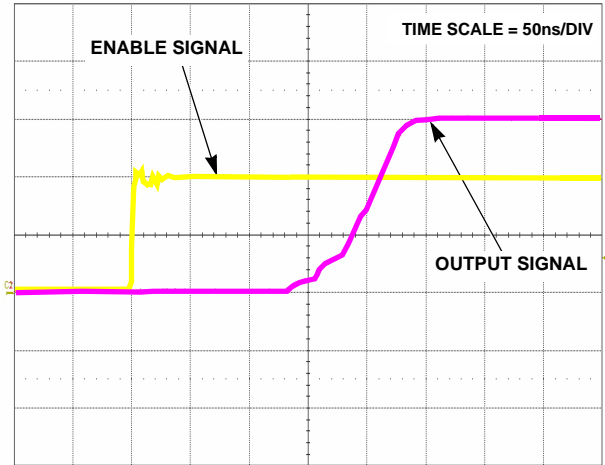


FIGURE 13. ENABLE TIME

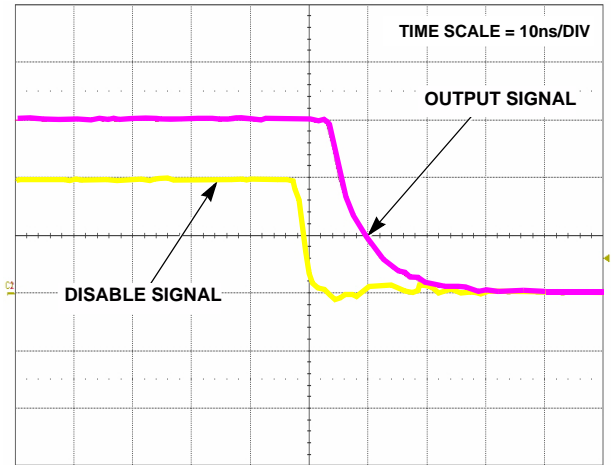


FIGURE 14. DISABLE TIME

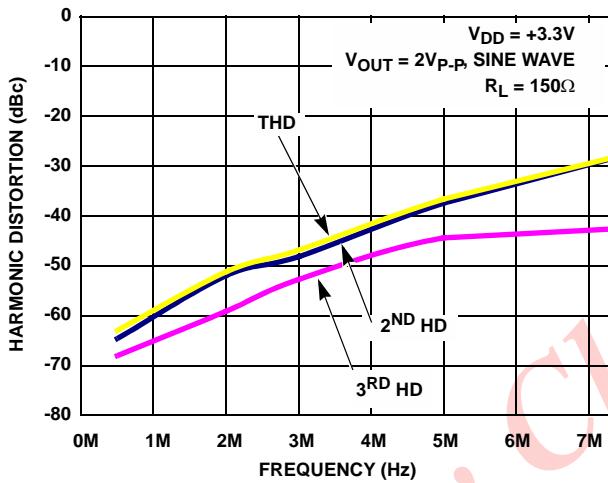


FIGURE 15. HARMONIC DISTORTION vs FREQUENCY

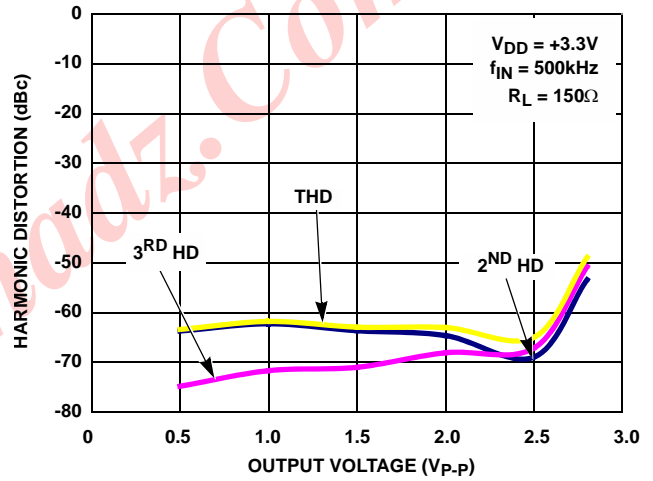


FIGURE 16. HARMONIC DISTORTION vs OUTPUT VOLTAGE

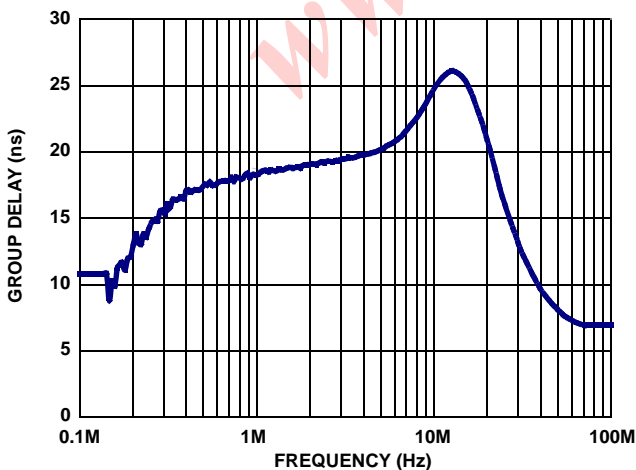


FIGURE 17. GROUP DELAY vs FREQUENCY

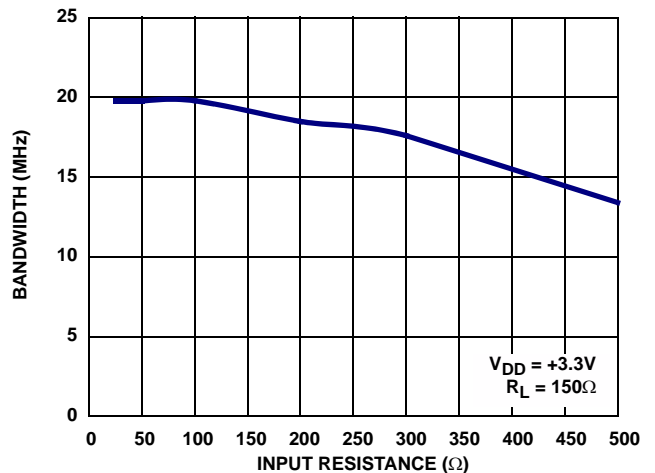


FIGURE 18. -3dB BANDWIDTH vs INPUT RESISTANCE

Typical Performance Curves $V_{DD} = +3.3V$, $R_L = 150\Omega$, unless otherwise noted

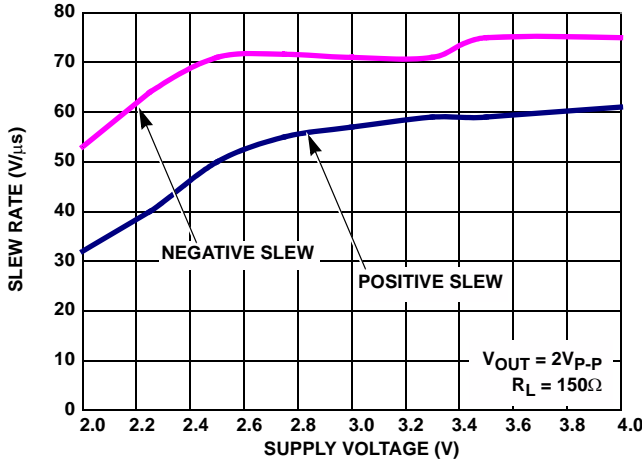


FIGURE 19. SLEW RATE vs SUPPLY VOLTAGE

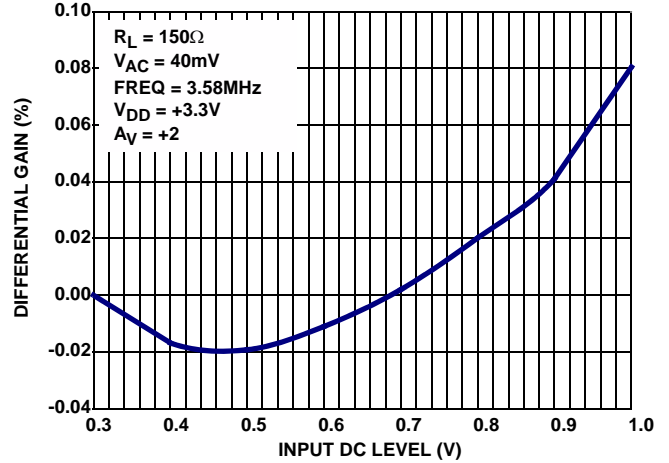


FIGURE 20. DIFFERENTIAL GAIN

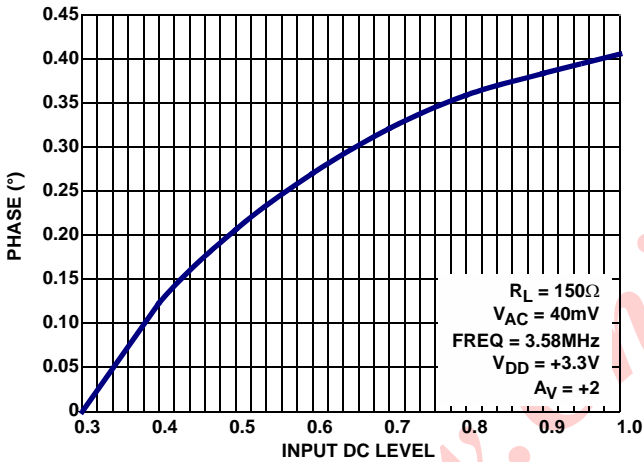


FIGURE 21. DIFFERENTIAL PHASE

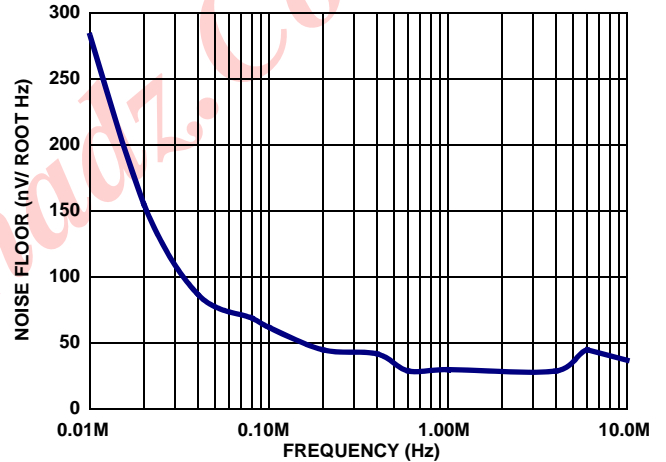


FIGURE 22. UNWEIGHTED NOISE FLOOR

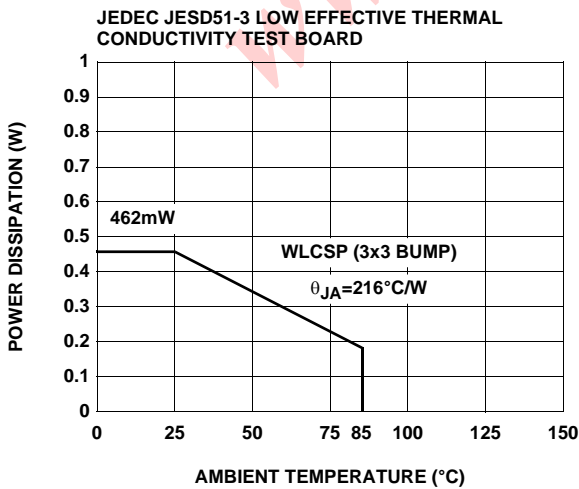


FIGURE 23. PACKAGE POWER DISSIPATION vs AMBIENT TEMPERATURE

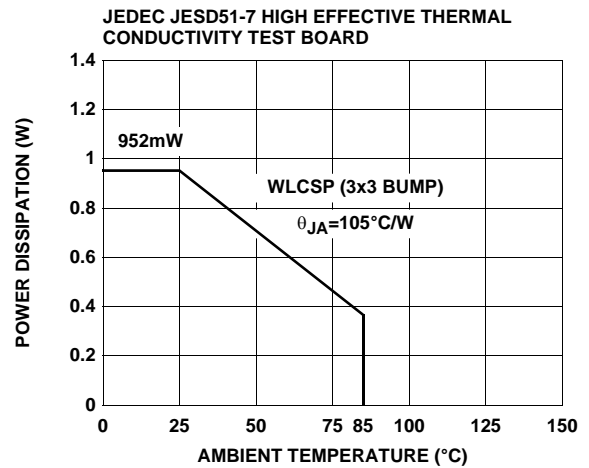


FIGURE 24. PACKAGE POWER DISSIPATION vs AMBIENT TEMPERATURE

Application Information

The ISL59121 is a single-supply rail-to-rail triple video low-pass filter and amplifier with a -3dB bandwidth of 18MHz. It provides anti-aliasing for component, s-video, and composite video signals. Its small size and low power make the ISL59121 ideal for portable video applications.

The Sallen Key Low Pass Filter

The Sallen Key is a classic low pass configuration. This provides a very stable low pass function, and in the case of the ISL59121, a three-pole roll-off at 18MHz. The three-pole function is accomplished with an RC low pass network placed in series with and before the Sallen Key. The first pole is formed by an RC network (including the impedance of the source driving the ISL59121), with poles two and three generated by a Sallen Key, creating a three-pole roll-off characteristic at 18MHz.

Output Coupling

The ISL59121 can be AC or DC coupled to its output. When AC coupled, a 220µF coupling capacitor is recommended to ensure that low frequencies are passed, preventing video "tilt" or "droop" across a line.

Output Drive Capability

The ISL59121 does not have internal short circuit protection circuitry. If the output is shorted indefinitely, the power dissipation could easily overheat the die or the current could eventually compromise metal integrity. Maximum reliability is maintained if the output current never exceeds ±40mA. This limit is set by the design of the internal metal interconnect. Note that for transient short circuits, the part is robust.

Short circuit protection can be provided externally with a back match resistor in series with the output placed close as possible to the output pin. In video applications this would be a 75Ω resistor and will provide adequate short circuit protection to the device. Care should still be taken not to stress the device with a short at the output.

Power Dissipation

With the high output drive capability of the ISL59121, it is possible to exceed the +125°C absolute maximum junction temperature under certain load current conditions. Therefore, it is important to calculate the maximum junction temperature for an application to determine if load conditions or package types need to be modified to assure operation of the amplifier in a safe operating area.

The maximum power dissipation allowed in a package is determined according to Equation 1:

$$PD_{MAX} = \frac{T_{JMAX} - T_{AMAX}}{\theta_{JA}} \quad (\text{EQ. 1})$$

Where:

T_{JMAX} = Maximum junction temperature

T_{AMAX} = Maximum ambient temperature

θ_{JA} = Thermal resistance of the package

The maximum power dissipation actually produced by an IC is the total quiescent supply current times the total power supply voltage, plus the power in the IC due to the load, or:

for sourcing:

$$PD_{MAX} = V_S \times I_{SMAX} + (V_S - V_{OUT}) \times \frac{V_{OUT}}{R_L} \quad (\text{EQ. 2})$$

for sinking:

$$PD_{MAX} = V_S \times I_{SMAX} + (V_{OUT} - V_S) \times I_{LOAD} \quad (\text{EQ. 3})$$

Where:

V_S = Supply voltage

I_{SMAX} = Maximum quiescent supply current

V_{OUT} = Maximum output voltage of the application

R_{LOAD} = Load resistance tied to ground

I_{LOAD} = Load current

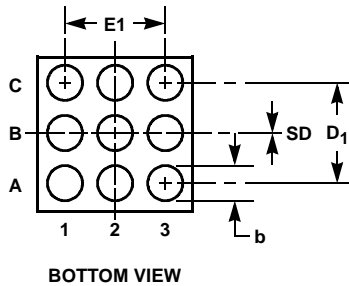
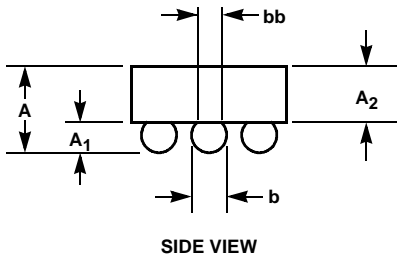
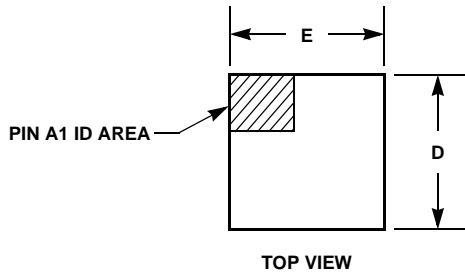
Power Supply Bypassing Printed Circuit Board Layout

As with any modern operational amplifier, a good printed circuit board layout is necessary for optimum performance. Lead lengths should be as short as possible. The power supply pin must be well bypassed to reduce the risk of oscillation. For normal single supply operation, a single 4.7µF tantalum capacitor in parallel with a 0.1µF ceramic capacitor from V_{DD} to GND will suffice.

Printed Circuit Board Layout

For good AC performance, parasitic capacitance should be kept to minimum. Use of wire wound resistors should be avoided because of their additional series inductance. Use of sockets should also be avoided if possible. Sockets add parasitic inductance and capacitance that can result in compromised performance.

Wafer Level Chip Scale Package (WLCSP)



W3x3.9B

3x3 ARRAY 9 BALL WAFER LEVEL CHIP SCALE PACKAGE (Intersil Standard)

| SYMBOL | MILLIMETERS | NOTES |
|----------------|--------------------|-------|
| A | 0.54 Min, 0.65 Max | - |
| A ₁ | 0.24 ±0.03 | - |
| A ₂ | 0.355 ±0.03 | - |
| b | 0.32 ±0.03 | - |
| bb | 0.30 REF. | - |
| D | 1.45 ±0.05 | - |
| D ₁ | 1.00 BASIC | - |
| E | 1.45 ±0.05 | - |
| E ₁ | 1.00 BASIC | - |
| e | 0.50 BASIC | - |
| SD | 0.00 BASIC | - |
| N | 9 | 3 |

Rev. 0 6/06

NOTES:

1. Dimensions are in Millimeters.
2. Dimensioning and tolerancing conform to ASME 14.5M-1994.
3. Symbol "N" is the actual number of solder balls.
4. Reference JEDEC MO-211-C, variation DD.

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